## Description

The F2910 is a high reliability, low insertion loss, $50 \Omega$ absorptive SPST RF switch designed for a multitude of wireless and RF applications. This device covers a broad frequency range from 30 MHz to 8000 MHz . In addition to providing low insertion loss, the F2910 also delivers excellent linearity and isolation performance while providing a $50 \Omega$ termination on RF2 in the isolation mode. The F2910 includes a patent pending constant impedance $K_{[\mid]}$feature for the RF2 port. $K_{[\mid]}$maintains near constant impedance when switching RF ports and improves hot switching ruggedness. $\mathrm{K}_{[\mid]}$minimizes VSWR transients and reduces phase and amplitude variations when switching.

The F2910 uses a single positive supply voltage supporting either 3.3 V or 1.8 V control logic.

## Competitive Advantage

The F2910 provides constant impedance for one RF port during transitions, improving a system's hot-switching ruggedness. The device also supports high power handling and high isolation.

- Constant impedance $\mathrm{K}_{\mathbf{| z |}}$ during switching transition
- Low insertion loss
- High isolation
- Excellent linearity
- Extended temperature range: $-55^{\circ} \mathrm{C}$ to $+105^{\circ} \mathrm{C}$


## Typical Applications

- Base Station 2G, 3G, 4G
- Portable Wireless
- Repeaters and E911 systems
- Digital Pre-Distortion
- Public Safety Infrastructure
- WIMAX Receivers and Transmitters
- Military Systems, JTRS radios
- RFID handheld and portable readers
- Cable Infrastructure
- Wireless LAN
- Test/ATE Equipment


## Features

- Insertion Loss

$$
\checkmark 0.58 \mathrm{~dB} \text { at } 2 \mathrm{GHz}
$$

- High Isolation
$\checkmark 51 \mathrm{~dB}$ at 2 GHz
- High Linearity
$\checkmark$ IIP3 of 65 dBm
- Wide Single Positive Supply Voltage Range
- 3.3 V and 1.8 V compatible control logic
- Operating temperature $-55^{\circ} \mathrm{C}$ to $+105^{\circ} \mathrm{C}$
- $2 \mathrm{~mm} \times 2 \mathrm{~mm} 8$-pin DFN package


## Block Diagram

Figure 1. Block Diagram


## Pin Assignments

Figure 2. Pin Assignments for $2 \mathrm{~mm} \times 2 \mathrm{~mm} \times 0.9 \mathrm{~mm}$ 8-VFQFP-N - Top View


## Pin Descriptions

Table 1. Pin Descriptions

| Number | Name | Description |
| :---: | :---: | :--- |
| 1,4 | NC | This pin may be connected to the paddle and can be grounded. |
| 2,3 | GND | Ground. Also, internally connected to the ground paddle. Ground this pin as close to the device as <br> possible. |
| 5 | RF1 | RF1 Port. Matched to $50 \Omega$ in the insertion loss state only. If this pin is not 0 V DC, then an external <br> coupling capacitor must be used. |
| 6 | $V_{D D}$ | Power Supply. Bypass to GND with capacitors shown in the Typical Application Circuit as close as <br> possible to pin. |
| 7 | V1 | Logic control pin. See Table 6 for proper logic setting. |
| 8 | EP | RF2 Port. Matched to 50 $\Omega$. If this pin is not 0V DC, then an external coupling capacitor must be used. |
|  | Exposed Pad. Internally connected to GND. Solder this exposed pad to a PCB pad that uses multiple <br> ground vias to provide heat transfer out of the device and into the PCB ground planes. These multiple <br> ground vias are also required to achieve the specified RF performance. |  |

## Renesns

## Absolute Maximum Ratings

Stresses beyond those listed below may cause permanent damage to the device. Functional operation of the device at these or any other conditions beyond those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Table 2. Absolute Maximum Ratings

| Parameter | Symbol | Minimum | Maximum | Units |
| :---: | :---: | :---: | :---: | :---: |
| V ${ }_{\text {D }}$ to GND | $V_{D D}$ | -0.3 | +6.0 | V |
| V1 to GND | $V_{\text {Logic }}$ | -0.3 | Lower of $\left(\mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V}, 3.6 \mathrm{~V}\right)$ | V |
| RF1, RF2 to GND | $\mathrm{V}_{\text {RF }}$ | -0.3 | +0.3 | V |
| RF Input Power Port 1 or 2 Other port terminated into $50 \Omega$ [a] | $\mathrm{P}_{\text {RF12 }}$ |  | 33 | dBm |
| RF Input Power Port 1 in isolation Port 2 terminated into $50 \Omega$ [a] | $\mathrm{P}_{\text {RF1_Iso }}$ |  | 23 |  |
| RF Input Power Port 2 in isolation Port 1 terminated into $50 \Omega$ [a] | $\mathrm{P}_{\text {RF2_ISO }}$ |  | 30 |  |
| Maximum Junction Temperature | $\mathrm{T}_{\text {max }}$ |  | +140 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature Range | T ${ }_{\text {st }}$ | -65 | +150 | ${ }^{\circ} \mathrm{C}$ |
| Lead Temperature (soldering, 10s) | TLEAD |  | +260 | ${ }^{\circ} \mathrm{C}$ |
| ElectroStatic Discharge - HBM (JEDEC/ESDA JS-001-2012) | $V_{\text {Esbhbm }}$ |  | $\begin{gathered} 2000 \\ \text { (Class 2) } \\ \hline \end{gathered}$ | V |
| ElectroStatic Discharge -CDM (JEDEC 22-C101F) | $V_{\text {Esdcom }}$ |  | $\begin{gathered} 1000 \\ \text { (Class C3) } \end{gathered}$ | V |

a. $\mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to $5.5 \mathrm{~V}, 30 \mathrm{MHz} \leq \mathrm{F}_{\mathrm{RF}} \leq 8000 \mathrm{MHz}, \mathrm{T}_{\mathrm{c}}=105^{\circ} \mathrm{C}, \mathrm{Z}_{\mathrm{S}}=\mathrm{Z}_{\mathrm{L}}=50$ ohms.

## Recommended Operating Conditions

Table 3. Recommended Operating Conditions

| Parameter | Symbol | Condition |  | Minimum | Typical | Maximum | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power supply voltage | $V_{\text {DD }}$ |  |  | 2.7 |  | 5.5 | V |
| Logic Input High Threshold | $\mathrm{V}_{\mathrm{IH}}$ | $2.7 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq 5.5 \mathrm{~V}$ |  | $1.1{ }^{\text {[a] }}$ |  | Lower of (VDD, 3.6) | V |
| Logic Input Low Threshold | $\mathrm{V}_{\text {IL }}$ |  |  | -0.3 [b] |  | 0.6 | V |
| Operating Temperature Range | TCASE | Exposed Paddle Temperature |  | -55 |  | +105 | ${ }^{\circ} \mathrm{C}$ |
| RF Frequency Range | $\mathrm{F}_{\mathrm{RF}}$ |  |  | 30 |  | 8000 [c] | MHz |
| RF Continuous Input CW Power (Non-Switched) [d] | PRF | RF1 or RF2 as the input (Insertion loss state) | $\mathrm{T}_{\mathrm{c}}=85^{\circ} \mathrm{C}$ |  |  | 30 | dBm |
|  |  |  | $\mathrm{T}_{\mathrm{c}}=105^{\circ} \mathrm{C}$ |  |  | 27 |  |
|  |  | RF1 as the input (Isolation state) | $\mathrm{T}_{\mathrm{c}}=85^{\circ} \mathrm{C}$ |  |  | 20 |  |
|  |  |  | $\mathrm{T}_{\mathrm{c}}=10{ }^{\circ} \mathrm{C}$ |  |  | 17 |  |
|  |  | RF2 as the input (Isolation state) | $\mathrm{T}_{\mathrm{c}}=85^{\circ} \mathrm{C}$ |  |  | 27 |  |
|  |  |  | $\mathrm{T}_{\mathrm{c}}=10{ }^{\circ} \mathrm{C}$ |  |  | 24 |  |
| RF Continuous Input Power (RF Hot Switching CW) ${ }^{[d]}$ | Prfsw | Applied to RF2 input switching between Insertion loss to Isolation states | $\mathrm{T}_{\mathrm{C}}=85^{\circ} \mathrm{C}$ |  |  | 24 | dBm |
|  |  |  | $\mathrm{T}_{\mathrm{C}}=105^{\circ} \mathrm{C}$ |  |  | 21 |  |
| RF1/2 Port Impedance | $Z_{\text {RFx }}$ | Insertion loss state |  |  | 50 |  | $\Omega$ |
| RF2 Port Impedance | $\mathrm{Z}_{\text {RFx }}$ | Isolation state |  |  | 50 |  | $\Omega$ |

a. Items in min/max columns in bold italics are Guaranteed by Test.
b. Items in min/max columns that are not bold/italics are Guaranteed by Design Characterization.
c. To achieve best performance from $5-8 \mathrm{GHz}$, the use of bypass capacitors as described in the Applications Circuit section is required.
d. Levels based on: $\mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to $5.5 \mathrm{~V}, 30 \mathrm{MHz} \leq \mathrm{F}_{\mathrm{RF}} \leq 8000 \mathrm{MHz}, \mathrm{Z}_{\mathrm{S}}=\mathrm{Z}_{\mathrm{L}}=50$ ohms. See Figure 3 for power handling derating vs RF frequency.

Figure 3. Maximum RF Input Operating Power vs. RF Frequency


## Renesns

## Electrical Characteristics

## Table 4. Electrical Characteristics

Typical Application Circuit: $V_{D D}=3.3 \mathrm{~V}, \mathrm{~T}_{\text {CASE }}=+25^{\circ} \mathrm{C}, \mathrm{F}_{\mathrm{RF}}=2 \mathrm{GHz}$, Driven Port $=\mathrm{RF} 2$, Input power $=0 \mathrm{dBm}, \mathrm{Z}_{\mathrm{S}}=\mathrm{Z}_{\mathrm{L}}=50 \Omega$. PCB board trace and connector losses are de-embedded unless otherwise noted. IIP2 / IIP3: $\mathrm{P}_{\mathrm{iN}}=13 \mathrm{dBm} /$ tone, 50 MHz spacing. Performance beyond 5 GHz based on application circuit (Figure 20) using best RF PCB design practices. See note c for details.

| Parameter | Symbol | Condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Logic Current | $\mathrm{IHH}_{\text {I }} \mathrm{IL}$ |  | -1 |  | +1 | $\mu \mathrm{A}$ |
| DC Current | lod | $\mathrm{V}_{\text {DD }}=3.3 \mathrm{~V}$ |  | 190 | 304 [a] | $\mu \mathrm{A}$ |
|  |  | $\mathrm{V}_{\text {DD }}=5.0 \mathrm{~V}$ |  | 220 | 374 |  |
| Insertion Loss | IL | 0.03 GHz |  | 0.38 |  | dB |
|  |  | 0.35 GHz |  | 0.44 |  |  |
|  |  | 1.0 GHz |  | 0.50 | 0.70 [b] |  |
|  |  | 2.0 GHz |  | 0.55 | 0.80 |  |
|  |  | 3.0 GHz |  | 0.60 | 0.85 |  |
|  |  | 4.0 GHz |  | 0.67 | 0.90 |  |
|  |  | 5.0 GHz |  | 0.75 | 1.00 |  |
|  |  | 6.0 GHz |  | 0.80 [c] |  |  |
|  |  | 7.0 GHz |  | 1.00 [c] |  |  |
|  |  | 8.0 GHz |  | 1.55 [c] |  |  |
| Isolation | ISO | 0.03 GHz |  | 85 |  | dB |
|  |  | 0.35 GHz | 66 | 73 |  |  |
|  |  | 1.0 GHz | 55 | 61 |  |  |
|  |  | 2.0 GHz | 45 | 51 |  |  |
|  |  | 3.0 GHz | 40 | 46 |  |  |
|  |  | 4.0 GHz | 35 | 41 |  |  |
|  |  | 5.0 GHz | 30 | 37 |  |  |
|  |  | 6.0 GHz |  | 33 [c] |  |  |
|  |  | 7.0 GHz |  | 29 [c] |  |  |
|  |  | 8.0 GHz |  | 26 [c] |  |  |
| Max RF2 Port VSWR During Switching | VSWR | Insertion Loss to Isolation |  | 3.3:1 |  |  |
|  |  | Isolation to Insertion Loss |  | 2.0:1 |  |  |
| RF1, RF2 Return Loss (Insertion Loss State) | RFRL | 2.0 GHz |  | 27 |  | dB |
|  |  | 3.0 GHz |  | 25 |  |  |
|  |  | 4.0 GHz |  | 20 |  |  |
|  |  | 5.0 GHz |  | 18 |  |  |
|  |  | 6.0 GHz |  | 20 [c] |  |  |
|  |  | 7.0 GHz |  | 25 [c] |  |  |
|  |  | 8.0 GHz |  | 13 [c] |  |  |

a. Items in min/max columns in bold italics are Guaranteed by Test.
b. Items in min/max columns that are not bold/italics are Guaranteed by Design Characterization.
c. To achieve performance beyond 5 GHz , the use of bypass capacitors ( $\mathrm{BOM} \mathrm{C} 2, \mathrm{C} 3$, and C 5 ) installed close to the device as embodied in the evaluation board per the application circuit (Figure 20) is required. See the appropriate Typical Operating Conditions graphs.

## Renesns

## Electrical Characteristics (Cont.)

Typical Application Circuit: $\mathrm{V}_{\mathrm{DD}}=3.3 \mathrm{~V}, \mathrm{~T}_{\mathrm{CASE}}=+25^{\circ} \mathrm{C}, \mathrm{F}_{\mathrm{RF}}=2 \mathrm{GHz}$, Driven Port $=\mathrm{RF} 2$, Input power $=0 \mathrm{dBm}, \mathrm{Z}_{\mathrm{S}}=\mathrm{Z}_{\mathrm{L}}=50 \Omega$. PCB board trace and connector losses are de-embedded unless otherwise noted. IIP2 / IIP3: $P_{\text {IN }}=13 \mathrm{dBm} /$ tone, 50 MHz spacing. Performance beyond 5 GHz based on application circuit (Figure 20) using best RF PCB design practices. See note c for details.

| Parameter | Symbol | Condition |  | Minimum | Typical | Maximum | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RF2 Return Loss (Isolation State) | RF RLLT | 2.0 GHz |  |  | 27 |  | dB |
|  |  | 3.0 GHz |  |  | 27 |  |  |
|  |  | 4.0 GHz |  |  | 25 |  |  |
|  |  | 5.0 GHz |  |  | 20 |  |  |
|  |  | 6.0 GHz |  |  | 15 [c] |  |  |
|  |  | 7.0 GHz |  |  | 12 [c] |  |  |
|  |  | 8.0 GHz |  |  | 10 [c] |  |  |
| Input 1dB Compression [d] | ICP ${ }_{\text {1ов }}$ | 0.03 GHz |  |  | 34 |  | dBm |
|  |  | 3.0 GHz |  |  | 35 |  |  |
|  |  | 4.0 GHz |  |  | 35 |  |  |
| Input 0.1dB Compression [d] | $\mathrm{ICP}_{0.1 \text { dв }}$ | 0.03 GHz |  |  | 33 |  | dBm |
|  |  | 3.0 GHz |  |  | 34 |  |  |
|  |  | 4.0 GHz |  |  | 34 |  |  |
| Input IP2 [e] | IIP2 | $\mathrm{F} 1=0.35 \mathrm{GHz} \mathrm{F} 2=0.40 \mathrm{GHz}$ |  |  | 123 |  | dBm |
|  |  | $\mathrm{F} 1=0.95 \mathrm{GHz} \mathrm{F} 2=1.00 \mathrm{GHz}$ |  |  | 124 |  |  |
|  |  | F1 $=2.40 \mathrm{GHz} \mathrm{F} 2=2.45 \mathrm{GHz}$ |  |  | 118 |  |  |
| Input IP3 [e] | IIP3 | 0.03 GHz |  |  | 65 |  | dBm |
|  |  | 0.35 GHz |  |  | 65 |  |  |
|  |  | 1.00 GHz |  |  | 68 |  |  |
|  |  | 2.40 GHz |  |  | 67 |  |  |
| Non-RF Driven Spurious [f] | Spurmax | Any port when externally terminated into $50 \Omega$ |  |  | -102 |  | dBm |
| Switching Time ${ }^{\text {a] }}$ | Tsw | 50\% control to 90\% RF |  |  | 265 | 500 | ns |
|  |  | 50\% control to 10\% RF |  |  | 225 | 500 |  |
|  |  | $50 \%$ control to RF settled to within +/- 0.1 dB of insertion loss value |  |  | 280 |  |  |
| Maximum Switching Rate | SW Rate |  |  |  | 25 |  | kHz |
| deo Feed-through |  | Peak transients during switching. | Rise Time |  | 25 |  |  |
| on RF Ports | V |  | Fall Time |  | 45 |  | pp |

a. Items in min/max columns in bold italics are Guaranteed by Test.
b. Items in min/max columns that are not bold/italics are Guaranteed by Design Characterization.
c. To achieve performance beyond 5 GHz , the use of bypass capacitors (BOM C2, C3, and C5) installed close to the device as embodied in the evaluation board per the application circuit (Figure 20) is required. See the appropriate Typical Operating Conditions graphs.
d. The input 0.1 and 1 dB compression point is a linearity figure of merit. Refer to Absolute Maximum Ratings section for the maximum RF input power.
e. RF1 or RF2 driven IIP2 / IIP3 results when in insertion loss state. IP2 Frequency = F1 + F2.
f. Spurious due to on-chip negative voltage generator. Spurious fundamental is approximately 5.7 MHz .
g. $F_{R F}=1 \mathrm{GHz}$.

## Thermal Characteristics

Table 5. Package Thermal Characteristics

| Parameter | Symbol | Value | Units |
| :--- | :---: | :---: | :---: |
| Junction to Ambient Thermal Resistance. | $\theta_{\mathrm{JA}}$ | 159.5 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| Junction to Case Thermal Resistance. <br> (Case is defined as the exposed paddle) | $\theta_{\mathrm{JC}}$ | 15.1 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| Moisture Sensitivity Rating (Per J-STD-020) |  | MSL 1 |  |

## Typical Operating Conditions (TOC)

Unless otherwise noted:

- $V_{D D}=3.3 \mathrm{~V}$.
- $Z_{L}=Z_{S}=50$ Ohms Single Ended.
- $F_{R F}=2 \mathrm{GHz}$.
- $P_{\text {IN }}=13 \mathrm{dBm}$ / tone applied to RF2 port for two tone linearity tests.
- Two tone frequency spacing $=50 \mathrm{MHz}$.
- All temperatures are referenced to the exposed paddle.
- Evaluation Kit traces and connector losses are de-embedded.
- Performance beyond 5 GHz as listed in the Electrical Characteristics is based on the application circuit (Figure 20) with bypass capacitors (BOM C2, C3, and C5) installed. The capacitors must be installed in close proximity to the device as embodied in the evaluation board with best practices followed for PCB design. Performance above 5 GHz de-rated as shown in Typical Performance Characteristics plots Figure 10 to Figure 13 when application circuit with bypass capacitors is not utilized.
- Unless otherwise noted, C2, C3 and C5 are installed in following plots.


## Typical Performance Characteristics

Figure 4. Insertion Loss vs. Frequency over Temperature


Figure 6. Return Loss vs. Frequency over Temp [RF1 Insertion Loss State]


Figure 8. Return Loss vs. Frequency over Temperature [RF2 Terminated State]


Figure 5. Isolation vs. Frequency over Temperature


Figure 7. Return Loss vs. Frequency over Temp [RF2 Insertion Loss State]


Figure 9. Evaluation Board Loss vs. Frequency


## Typical Performance Characteristics

Figure 10. Insertion Loss vs. Frequency With and Without Capacitors


Figure 12. Return Loss vs. Frequency With and Without Capacitors


Figure 14. Input IP3 vs. Frequency


Figure 11. Isolation vs. Frequency With and Without Capacitors


Figure 13. Return Loss vs. Frequency With and Without Capacitors [State 0]


Figure 15. 1 dB Compression at $\mathbf{3} \mathbf{~ G H z}$


## Typical Performance Characteristics

Figure 16. Switching Time Isolation to Insertion loss


Figure 17. Switching Time Insertion Loss to Isolation


## Evaluation Kit Picture

Figure 18. Top View


Figure 19. Bottom View


## Control Mode

Table 6. Switch Control Truth Table

| V1 <br> (Logic) | State | Port Match |
| :---: | :---: | :---: |
| 0 | Isolation | RF1 port reflective, RF2 port terminated into 50 ohms |
| 1 | Insertion Loss | RF1 and RF2 port matched to 50 ohm |

## Evaluation Kit / Applications Circuit

## Figure 20. Electrical Schematic

Note: The use of bypass capacitors $\mathrm{C} 2, \mathrm{C} 3$, and C 5 as listed in the BOM (Table 7) is required to achieve performance as listed in the Electrical Characteristics for frequencies beyond 5 GHz . The capacitors must be installed in close proximity to the device as embodied in the evaluation board with best practices followed for PCB design.


Table 7. Bill of Material (BOM)

| Part Reference | QTY | Description | Manufacturer Part \# | Manufacturer |
| :---: | :---: | :---: | :---: | :---: |
| C1 | 1 | $0.1 \mu \mathrm{~F} \pm 10 \%, 16 \mathrm{~V}, \mathrm{X} 7 \mathrm{R}$, Ceramic Capacitor (0402) | GRM155R71C104KA88D | Murata |
| C2 | 1 | $0.5 \mathrm{pF} \pm 0.1 \mathrm{pF}, 50 \mathrm{~V}, \mathrm{COG}$, Ceramic Capacitor (0402) | GJM1555C1HR50BB01 | Murata |
| C3 | 1 | $4.5 \mathrm{pF} \pm 0.1 \mathrm{pF}, 50 \mathrm{~V}, \mathrm{COG}$, Ceramic Capacitor (0402) | GJM1555C1H4R5BB01D | Murata |
| C4 | 0 | Not Installed (0402) |  |  |
| C5 | 1 | $4.9 \mathrm{pF} \pm 0.1 \mathrm{pF}, 50 \mathrm{~V}, \mathrm{COG}$, Ceramic Capacitor (0402) | GJM1555C1H4R9BB01 | Murata |
| R1 | 1 | $15 \mathrm{k} \Omega \pm 1 \%, 1 / 10 \mathrm{~W}$, Resistor (0402) | ERJ-2RKF1502X | Panasonic |
| R2 | 1 | $18 \mathrm{k} \Omega \pm 1 \%, 1 / 10 \mathrm{~W}$, Resistor (0402) | ERJ-2RKF1802X | Panasonic |
| R3, R4 | 2 | $0 \Omega 1 / 10 \mathrm{~W}$, Jumper (0402) | ERJ-2GE0R00X | Panasonic |
| J1-J4 | 4 | SMA Edge Mount | 142-0761-881 | Cinch Connectivity |
| J5 | 1 | CONN HEADER VERT 4x2 POS GOLD | 67997-108HLF | Amphenol FCI |
| TP1 | 0 | Not Installed (Red Test Point Loop) |  |  |
| TP2, TP3 | 0 | Not Installed (Black Test Point Loop) |  |  |
| U1 | 1 | SPST Switch $2 \mathrm{~mm} \times 2 \mathrm{~mm} 8$ pin DFN | F2910NBGP | IDT |
|  | 1 | Printed Circuit Board | F2910 EVKIT REV 01 | IDT |

## Evaluation Kit Operation

## Default Start-up

Control pins include no internal pull-down resistors to logic LOW or pull-up resistors to logic HIGH.

## Power Supplies

A common $V_{c c}$ power supply should be used for all pins requiring DC power. All supply pins should be bypassed with external capacitors to minimize noise and fast transients. Supply noise can degrade noise figure and fast transients can trigger ESD clamps and cause them to fail. Supply voltage change or transients should have a slew rate smaller than $1 \mathrm{~V} / 20 \mu \mathrm{~s}$. In addition, all control pins should remain at 0 V $( \pm 0.3 \mathrm{~V}$ ) while the supply voltage ramps or while it returns to zero.

## Control Pin Interface

If control signal integrity is a concern and clean signals cannot be guaranteed due to overshoot, undershoot, ringing, etc., the following circuit at the input of each control pin is recommended. This applies to control pin 7 as shown in Figure 21. If bypass capacitor C 5 as described in the application circuit (Figure 20) is used to achieve high frequency performance optimization, the use of an additional 2 pF capacitor as shown in Figure 21 is not necessary.

Figure 21. Control Pin signal integrity improvement circuit


## Renesas

## External Supply Setup

Set up a $V_{c c}$ power supply in the voltage range of 2.7 V to 5.5 V with the power supply output disabled.

## Logic Control Setup

Using the EVKIT to manually set the control logic:
On connector J5, connect a 2-pin shunt from $\mathrm{V}_{\mathrm{cc}}$ (pin 3) to $\mathrm{V}_{\text {Logic }}$ (pin 4). This connection provides the $\mathrm{V}_{\mathrm{cc}}$ voltage supply to the Eval Board logic control pull-up network. Resistors R 1 and R 2 form a voltage divider to set the $\mathrm{V}_{\mathbb{H}}$ level over the 2.7 V to $5.5 \mathrm{~V} \mathrm{~V}_{\mathrm{Cc}}$ range for manual logic control.

Connector J 5 has one logic input pin: V1 (pin 5). See Table 6 for Logic Truth Table. With the pull-up network enabled (as noted above) this pin can be left open to provide a logic high through pull-up resistor R1. To set a logic low for V1, connect a 2-pin shunt on J5 from VctL (pin 5) to GND (pin 6).

Note that when using the on board $\mathrm{R} 1 / \mathrm{R} 2$ voltage divider, the current draw from the $\mathrm{V}_{\mathrm{cc}}$ supply will be higher by approximately $\mathrm{V}_{\mathrm{cc}} / 33 \mathrm{k} \Omega$.
Using external control logic:
Pins $3,4,6,7$, and 8 of J 5 should have no external connections. External logic control is applied to J 5 V 1 (pin 5). See Table 5 for the Logic Truth Table.

## Turn On Procedure

Setup the supplies and EVKIT as noted in the External Supply Setup and Logic Control Setup sections above.

## Enable the Vcc supply.

Set the desired logic setting to achieve the desired Table 5 configuration. Note that external control logic should not be applied without $\mathrm{V}_{c c}$ being present.

## Turn Off Procedure

Set the logic control to a logic low.
Disable the $\mathrm{V}_{\mathrm{cc}}$ supply.

## Package Drawings

Figure 22. Package Outline Drawing


## Recommended Land Pattern

Figure 23. Recommended Land Pattern


## Ordering Information

| Orderable Part Number | Package | MSL Rating | Shipping Packaging | Temperature |
| :---: | :---: | :---: | :---: | :---: |
| F2910NBGP | $2 \times 2 \times 0.9 \mathrm{~mm} 8$-VFQFP-N | MSL1 | Bulk | $-55^{\circ}$ to $+105^{\circ} \mathrm{C}$ |
| F2910NBGP8 | $2 \times 2 \times 0.9 \mathrm{~mm}$ 8-VFQFP-N | MSL1 | Tape and Reel | $-55^{\circ}$ to $+105^{\circ} \mathrm{C}$ |
| F2910EVBI | Evaluation Board |  |  |  |

## Marking Diagram

1. Line 1 is the part number.
2. Line $2-$ " 6 " is last digit of the year.

2910
3. Line $2-$ " $U$ " is the workweek code

6UAG
4. Line 2 - AG is the sequential code

## Renesns

## Revision History

| Revision Date |  |
| :---: | :--- |
| $2016-$ Aug-29 | Initial Release |

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