

SANYO Semiconductors DATA SHEET

LV8760T — Forward/Reverse H-bridge Driver

Overview

The LV8760T is an H-bridge driver that can control four operation modes (forward, reverse, brake, and standby) of a motor. The low on-resistance, zero standby current, highly efficient IC is optimal for use in driving brushed DC motors for office equipment.

Features

- Forward/reverse H-bridge motor driver: 1 channel
- Built-in current limiter circuit
- Built-in thermal protection circuit
- Built-in short-circuit protection function

Specifications

Absolute Maximum Ratings at Ta = 25°C

Parameter Symbol		Conditions Ratings		Unit
Supply voltage	VM max		38	V
	V _{CC} max		6	V
Output peak current	I _O peak	tw ≤ 20ms, duty 5%	4	Α
Output continuous current I _O ma			3	Α
Logic input voltage	V _{IN}		-0.3 to V _{CC} +0.3	V
Allowable power dissipation Pd n		Mounted on a specified board. *	3.3	W
Operating temperature	Topr		-20 to +85	°C
Storage temperature	Tstg		-55 to +150	°C

^{*} Specified circuit board: 90mm×90mm×1.6mm, glass epoxy 2-layer board (2S0P), with backside mounting.

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Allowable Operating Ratings at $Ta = 25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage range	VM		9 to 35	V
	Vcc		3 to 5.5	V
VREF input voltage	VREF		0 to V _{CC} -1.8	V
Logic input voltage	V _{IN}		0 to V _{CC}	V

Electrical Characteristics at Ta = 25 °C, VM = 24V, $V_{CC} = 5V$, VREF = 1.5V

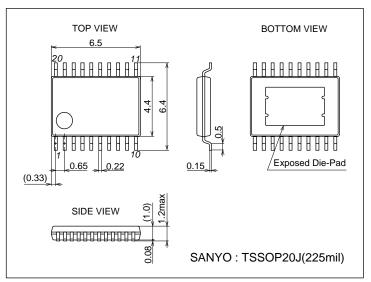
Parameter	Symbol	Conditions	Ratings			Unit	
raidinotor	Cymbol	Conditions	min	typ	max		
General							
Standby mode current drain 1	IMst	PS = "L"			1	μΑ	
Standby mode current drain 2	I _{CC} st	PS = "L"			1	μΑ	
Operating mode current drain 1	IM	PS = "H", IN1 = "H", with no load		1	1.3	mA	
Operating mode current drain 2	^I cc	PS = "H", IN1 = "H", with no load		3	4	mA	
VREG output voltage	VREG	$I_O = -1mA$	4.75	5	5.25	V	
V _{CC} low-voltage cutoff voltage	VthV _{CC}		2.5	2.7	2.9	V	
Low-voltage hysteresis voltage	VthHIS		120	150	180	mV	
Thermal shutdown temperature	TSD	Design guarantee *	155	170	185	°C	
Thermal hysteresis width	ΔTSD	Design guarantee *		40		°C	
Output block							
Output on resistance	Ron1	I _O = 3A, sink side		0.2	0.25	Ω	
	Ron2	I _O = -3A, source side		0.32	0.40	Ω	
Output leakage current	l _O leak	V _O = 35V			50	μΑ	
Rising time	tr	10% to 90%		200	500	ns	
Falling time	tf	90% to 10%		200	500	ns	
Input output delay time	tpLH	IN1 or IN2 to OUTA or OUTB (L \rightarrow H)		550	700	ns	
	tpHL	IN1 or IN2 to OUTA or OUTB (H \rightarrow L)		550	700	ns	
Charge pump block	•		'		'		
Step-up voltage	VGH	VM = 24V	28.0	28.7	29.8	V	
Rising time	tONG	VG = 0.1μF		250	500	μs	
Oscillation frequency	Fcp		115	140	165	kHz	
Control system input block	•		<u> </u>	l.	<u> </u>		
Logic pin input current 1	I _{IN} L	V _{IN} = 0.8V adaptive pin : PS	5.6	8	10.4	μΑ	
	I _{IN} H	V _{IN} = 5V adaptive pin : PS	56	80	104	μΑ	
Logic pin input current 2	I _{IN} L	V _{IN} = 0.8V adaptive pin : IN1, IN2	5.6	8	10.4	μΑ	
	I _{IN} H	V _{IN} = 5V adaptive pin : IN1, IN2	35	50	65	μΑ	
Logic pin input H-level voltage	V _{IN} H	adaptive pin : PS, IN1, IN2	2.0			V	
Logic pin input L-level voltage	V _{IN} L	adaptive pin : PS, IN1, IN2			0.8	V	
Current limiter block							
VREF input current	IREF		-0.5			μΑ	
Current limit comparator threshold voltage	Vthlim	VREF = 1.5V	0.285	0.3	0.315	V	
Short-circuit protection block	1	1	1				
SCP pin charge current	Iscp	SCP = 0V	3.5	5	6.5	μΑ	
Comparator threshold voltage	Vthscp		0.8	1	1.2	V	
	1						

^{*} Design guarantee value and no measurement is made.

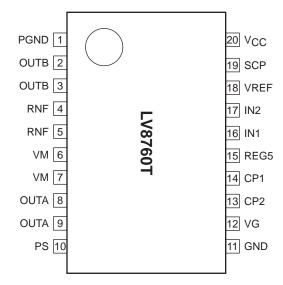
Package Dimensions

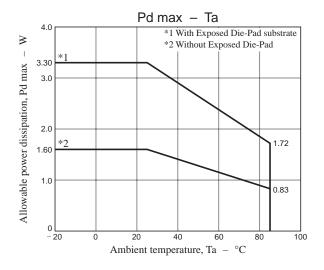
unit: mm (typ)

3279



Pin Assignment



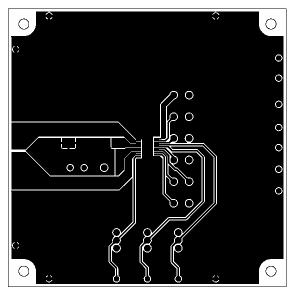


Substrate Specifications (Substrate recommended for operation of LV8760T)

Size : $90\text{mm} \times 90\text{mm} \times 1.6\text{mm}$ (two-layer substrate [2SOP])

Material : Glass epoxy

Copper wiring density : L1 = 95% / L2 = 95%



L1: Copper wiring pattern diagram

L2 : Copper wiring pattern diagram

Cautions

- 1) The data for the case with the Exposed Die-Pad substrate mounted shows the values when 90% or more of the Exposed Die-Pad is wet.
- 2) For the set design, employ the derating design with sufficient margin.

Stresses to be derated include the voltage, current, junction temperature, power loss, and mechanical stresses such as vibration, impact, and tension.

Accordingly, the design must ensure these stresses to be as low or small as possible.

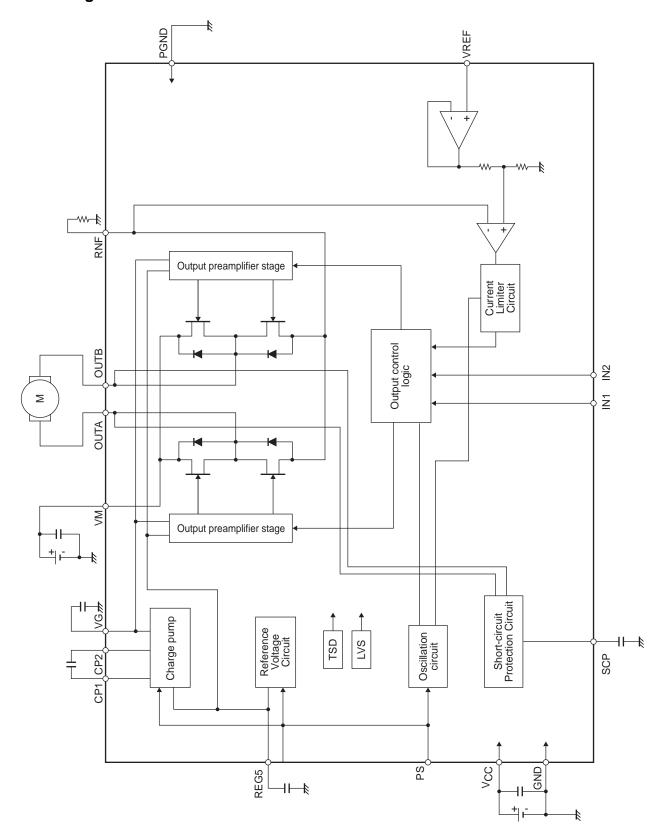
The guideline for ordinary derating is shown below:

- (1)Maximum value 80% or less for the voltage rating
- (2)Maximum value 80% or less for the current rating
- (3)Maximum value 80% or less for the temperature rating
- 3) After the set design, be sure to verify the design with the actual product.

Confirm the solder joint state and verify also the reliability of solder joint for the Exposed Die-Pad, etc.

Any void or deterioration, if observed in the solder joint of these parts, causes deteriorated thermal conduction, possibly resulting in thermal destruction of IC.

Block Diagram



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Pin Functions

	ctions		
Pin No.	Pin Name	Pin Functtion	Equivalent Circuit
16	IN1	Output control signal input pin 1.	Voc
17	IN2	Output control signal input pin 2.	Vcc O
			→
			10kΩ
			100kΩ + 100kΩ
40	DO	Deven and signal involves	GNDO
10	PS	Power save signal input pin.	Vcc o
			50kΩ \(\beta\)
			10kΩ
			50kΩ ξ
			GND O
18	VREF	Reference voltage input pin for output current limit setting.	Vcco
			500Ω
			*
			GND O
19	SCP	Short-circiut protection circuit, detection time setting capacitor connection pin.	Vcc o
			500Ω
			GND O
20	VCC	Power supply connection pin for control block.	

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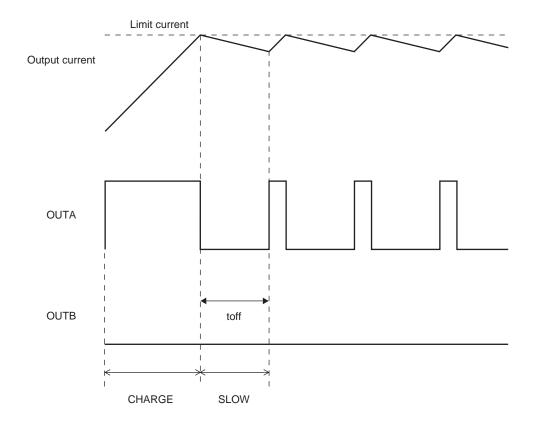
Pin No.	Pin Name	Pin Functtion	Equivalent Circuit
6, 7 8, 9 4, 5 2, 3	VM OUTA RNF OUTB PGND	Motor power-supply connection pin. OUTA output pin. Current sense resistor connection pin. OUTB output pin. Power ground.	REG5 0 8 9 500Ω 1) 45
14 13 12	CP1 CP2 VG	Charge pump capacitor connection pin. Charge pump capacitor connection pin. Charge pump capacitor connection pin.	GND 0 13 (2) REG5 0 100Ω GND 0 100Ω
15	REG5	Internal reference voltage output pin.	VM \circ $2k\Omega$ $25k\Omega$ GND \circ
11	GND	Ground.	

DC Motor Driver

1.DCM output control logic

Contol Input			Ou	tput	Mode	
PS	IN1	IN2	OUTA	OUTB	Mode	
L	*	*	OFF	OFF	Standby	
Н	L	L	OFF	OFF	Output OFF	
Н	Н	L	Н	L	CW (forward)	
Н	L	Н	L	Н	CCW (reverse)	
Н	Н	Н	L	L	Brake	

2. Current limit control timing chart



Braking operation time in current limit mode can be set by connecting a capacitor between SCP and GND pins. This setting is the same as the time setting required to turn off the outputs when an output short-circuit occurs as explained in the section entitled "Output Short-circuit Protection Function." See "Output Short-circuit Protection Function," for the setting procedure.

3. Setting the current limit value

The current limit value of the DCM driver is determined by the VREF voltage and the resistance (RNF) connected across the RNF and GND pins using the following formula :

Ilimit [A] =
$$(VREF [V] /5) /RNF [\Omega])$$

Assuming VREF = 1.5V, RNF = 0.2Ω , the current limit is :

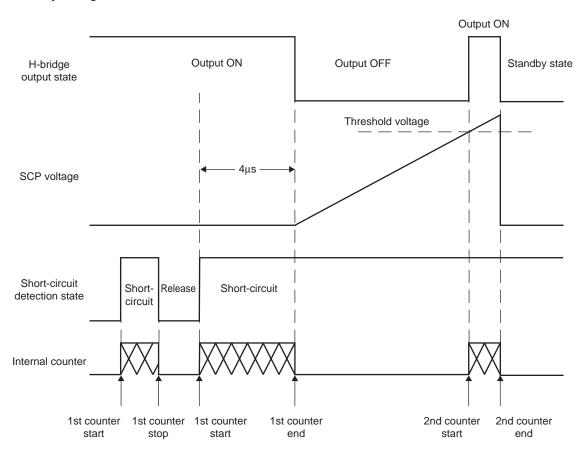
 $Ilimit = 1.5V/5/0.2\Omega = 1.5A$

Output short-circuit protection function

The LV8760T incorporates an output short-circuit protection circuit. It turns the ouputs off to prevent destruction of the IC if a problem such as an output pin being shorted to the motor power supply or ground occurs.

1. Protection function operation (Latch method)

The short-circuit protection circuit is activated when it detects the output short-circuit state. If the short-circuit state continues for the internally preset period ($\approx 4\mu s$), the protection circuit turns off the output from which the short-circuit state has been detected. Then it turns the output on again after a lapse of the timer latch time described later. If the short-circuit state is still detected, it changes all the outputs to the standby mode and retains the state. The latched state is released by setting the PS to L.



2. How to set the SCP pin constant (timer latch-up setting)

The user can set the time at which the outputs are turned off when a short-circuit occurs by connecting a capacitor across the SCP and GND pins. The value of the capacitor can be determined by the following formula:

Timer latch-up : Tocp $Tocp \approx C \times V/I[s]$

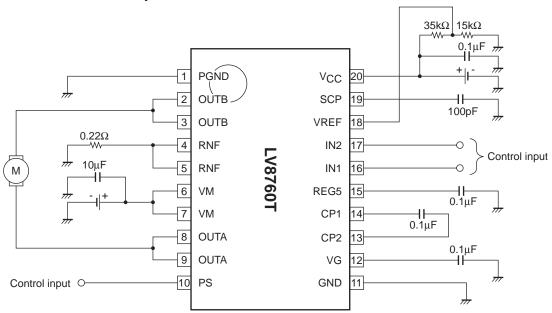
 $V: Comparator\ threshold\ voltage\ (1V\ typical)$

I : SCP charge current (5µA typical)

When a capacitor with a capacitance of 50pF is connected across the SCP and GND pins, for example, Tscp is calculated as follows :

$$Tscp = 50pF \times 1V/5\mu A = 10\mu s$$

Application Circuit Example



Setting the current limit value

When $V_{CC} = 5V$, $V_{ref} = 1.5V$ $Ilimit = V_{ref}/5/RNF$

 $= 1.5V/5/0.22\Omega = 1.36A$

Setting the current limit regeneration time and short-circuit detection time

Tscp
$$\approx$$
 C \times V/I
= 100pF \times 1V/5 μ A
= 20 μ s

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